

## IQ Impairments and Corrections in Ultra-wideband transmitters

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#### Outline

- 1. Motivations
- 2. Mathematical formulation
- 3. Image suppression impact on Comm. Systems.
- 4. Corrective Topologies
- 5. Adaptive blind estimations methods
- 6. Experimental data
- 7. Summary
- 8. Future work





#### Motivations



- 1. Multi-channel Transmitters
- 2. Ultra-Wideband Transmitters
  - a. Beam hopping and time slicing satellite communications
  - b. 5G waveforms





#### MATHEMATICAL FORMULATION AND PICTORIAL DEPICTIONS **IQ** IMPAIRMENTS

(3)

Modulated Baseband signal

$$y_m(t) = y_i(t) + jy_q(t)$$

Translated to RF

$$y_{trans}(t) = real\{y_m(t)e^{-jw_{lo}t}\}$$

I-component

 $y_i(t) = x_i(t)\cos(w_c t)$ 

Substitution and manipulation

![](_page_3_Figure_9.jpeg)

**I-Component** 

depicting an even-mode

$$Y_{i}(w) = F\{y_{i}(t)\} = F\{x_{i}(t)\} * F\{\cos(w_{c}t)\}$$
  
=  $X_{i}(w) \left\{\frac{e^{jw_{c}t} + e^{-jw_{c}t}}{2}\right\} Y_{i}(w)$   
=  $\frac{X_{i}(w)}{2} * [\delta(w - w_{c}) + \delta(w + w_{c})]$  (4)

![](_page_4_Picture_0.jpeg)

### MATHEMATICAL FORMULATION AND PICTORIAL DEPICTIONS IQ

Q-component

$$y_q(t) = -x_q(t)\sin(w_c t)$$
 (5)

Substitution and manipulation

$$Y_{q}(w) = F\{x_{q}(t)\} * F\{-\sin w_{c}t\}$$
  
=  $X_{q}(w) * F\{\frac{e^{-jw_{c}t}-e^{jw_{c}t}}{2j}\}$  (6)  
=  $\frac{X_{q}(w)}{2j} * [\delta(w-w_{c}) - \delta(w+w_{c})$ 

# Frequency Domain representation of Qcomponent depicting an odd-mode

![](_page_4_Picture_7.jpeg)

![](_page_5_Picture_0.jpeg)

#### MATHEMATICAL FORMULATION AND PICTORIAL DEPICTIONS IQ IMPAIRMENTS CONT.

Rewriting the translation equation

$$y_{trans}(t) = \frac{y_m(t)}{2} e^{-jw_{lo}t} + \frac{y_m^*(t)}{2} e^{jw_{lo}t} \quad (7)$$
Assuming,  $Y_m(w) = Y_m^*(w)$ 

$$= \frac{Y_m(w)}{2} * [\delta(w + w_{lo}) + \delta(w - w_{lo})] \quad (8)$$
Baseband
$$Ideal Direct-Conversion$$

$$Ideal Direct-Conversion$$

$$Ideal Direct-Conversion$$

$$Ideal Direct-Conversion$$

$$Ideal Direct-Conversion$$

$$Ideal Direct-Conversion$$

Adding I and Q to translation

Ideal direction conversion with baseband offset

$$Y_{m}(w) = \frac{Y_{i}(w)}{2} * \left[\delta(w - w_{c}) + \delta(w + w_{c})\right] + \frac{Y_{q}(w)}{2} * \left[\delta(w - w_{c}) - \delta(w + w_{c})\right]$$
(9)

![](_page_6_Figure_0.jpeg)

$$* [\delta(w + w_{lo}) + \delta(w - w_{lo})].$$

Wanted Signal

$$\frac{X_i(w) + X_q(w)}{4} * \delta(w - w_c) \quad (11) \quad \text{Vector relationship of I/Q Components}$$

Image Signal

$$\frac{X_i(w) - X_q(w)}{4} * \delta(w + w_c)$$
 (12)

![](_page_6_Picture_6.jpeg)

![](_page_7_Picture_0.jpeg)

#### IMPACT ON IMAGE-SUPPRESSION ON COMMUNICATION SYSTEMS

![](_page_7_Figure_2.jpeg)

Degradation of signal

$$CNIR = 10\log_{10}\left(\left(\frac{1}{\frac{SNR_{input}}{10}} + \frac{1}{\frac{CNIR_{image}}{10}}\right)^{-1}\right)$$
(13)

Rule of thumb: Keep CIR 15 dB below target SNR value

![](_page_7_Picture_6.jpeg)

![](_page_8_Picture_0.jpeg)

#### Frequency Impact on IQ imbalances

![](_page_8_Figure_2.jpeg)

Total Amplitude Imbalance

$$G_{imb}(w) = L_{imb}(w) * A_{imb}$$

**Total Phase Imbalance** 

$$\phi_{imb}(w) = \angle L_{imb}(w) + \theta_{imb} + \theta_{delay} \qquad (17)$$

Typical direct conversion transmitter implementation (16) with amplitude and phase imbalance.

![](_page_8_Picture_8.jpeg)

![](_page_9_Picture_0.jpeg)

#### **Corrective Topology Time Domain**

Imbalance Compensation

$$y_{i-corr}(t) = \alpha \{y_i(t) + \beta y_q(t)\} \quad (18)$$

Mostly Amplitude Compensation

$$\alpha = (1 + A_{imb})\cos(\theta_{imb})$$

Phase Compensation

 $\beta = sin(\theta_{imb})$ 

![](_page_9_Figure_8.jpeg)

Wideband corrective topology for phase and amplitude (20) imbalances for direct conversion transmitter

![](_page_9_Picture_10.jpeg)

![](_page_10_Picture_0.jpeg)

#### **Corrective Topology Time Domain**

![](_page_10_Figure_2.jpeg)

Constant slope FIR filter for frequency dependent imbalance correction.

![](_page_10_Figure_4.jpeg)

Phase imbalance over baseband frequency as a result of time delay imbalance

![](_page_10_Picture_6.jpeg)

![](_page_11_Picture_0.jpeg)

#### Adaptive Blind Estimation IQ IMPAIRMENTS

![](_page_11_Figure_2.jpeg)

![](_page_11_Picture_3.jpeg)

![](_page_12_Picture_0.jpeg)

#### Adaptive Blind Estimation IQ IMPAIRMENTS

Orthogonality

$$Error = \int_{-l}^{l} I(t)Q(t)dt,$$

Orthogonal error accumulation

$$\phi_{imb-new} = \phi_{imb-old} + \lambda_{phase} I(t) Q_{corr1}(t) \quad (2)$$

Amplitude error accumulation

$$G_{imb-new} = G_{imb-old} + \lambda_{gain} [I(t)^2 - Q_{corr2}(t)^2] \qquad (2$$

![](_page_12_Figure_8.jpeg)

Plot of sin and cos function to illustrate even and odd properties.

(23)

![](_page_12_Picture_11.jpeg)

![](_page_13_Picture_0.jpeg)

#### Adaptive Blind Estimation IQ IMPAIRMENTS

![](_page_13_Figure_2.jpeg)

$$Q_{corr2}(t) = G_{imb-new}Q_{corr1}(t).$$

Vector relationship of I/Q Components

image

![](_page_13_Picture_5.jpeg)

![](_page_14_Picture_0.jpeg)

#### **Experimental Results**

![](_page_14_Figure_2.jpeg)

Amplitude and phase correction of baseband IQ imbalances

![](_page_14_Figure_4.jpeg)

Image error vs gain/phase imbalance error

![](_page_14_Picture_6.jpeg)

![](_page_15_Picture_0.jpeg)

#### **Experimental Results**

![](_page_15_Figure_2.jpeg)

using gradient decent algorithm

![](_page_15_Figure_4.jpeg)

![](_page_15_Picture_5.jpeg)

![](_page_16_Picture_0.jpeg)

#### **Experimental Results**

Agilent Spectrum Analyzer - Swept SA												
<mark>.x</mark> Mar	ker 1	RF ▲ -24	50 Ω 0.000	AC 000000	MHz	SE	e Run	Avg Type	ALIGN AUTO : Log-Pwr	09:43:50 A TRAI	M May 01, 2018 CE <b>1 2 3 4 5 6</b> PE W Atanatata	Peak Search
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-30.0							X2					Next Pk Right
-40.0 -50.0												Next Pk Left
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Msg UFile <screen_0006.png> saved status</screen_0006.png>												

![](_page_16_Figure_3.jpeg)

![](_page_16_Picture_4.jpeg)

![](_page_17_Picture_0.jpeg)

#### Summary

- 1. Mathematically derived root cause of image
- 2. Explored implications on image suppression in communications
- 3. Identified frequency compensation methods for phase/gain imbalance
- 4. Blind estimation and compensation methods for dynamic systems presented
- 5. Experimental results for frequency and dynamic compensation methods presented

![](_page_17_Picture_7.jpeg)

![](_page_18_Picture_0.jpeg)

#### Future Work

- 1. Further exploration of different imbalance estimation methods
- 2. Compensation methods in frequency domain

![](_page_18_Picture_4.jpeg)

![](_page_19_Picture_0.jpeg)

#### Additional slides Linearity of estimation methods

![](_page_19_Figure_2.jpeg)

![](_page_20_Picture_0.jpeg)

#### Additional slides Linearity of estimation methods Cont.

![](_page_20_Figure_2.jpeg)

![](_page_20_Figure_3.jpeg)

![](_page_21_Picture_0.jpeg)

#### Additional Slides Corrective Topology Frequency Domain

![](_page_21_Figure_2.jpeg)

Wideband corrective topology for phase and amplitude imbalances for direct conversion transmitter

Fourier Transform of corrective domain structure

$$Y_{i-corr}(w) = y_{i-corr}(t) = \alpha F\{y_i(t)\} + \alpha \beta F\{y_q(t)\}$$
(26)

![](_page_21_Picture_6.jpeg)

![](_page_22_Picture_0.jpeg)

#### Additional Slides

#### Corrective Topology Frequency Domain Cont.

![](_page_22_Figure_3.jpeg)

Frequency selective corrective topology for phase and amplitude imbalances for direct conversion transmitter

![](_page_22_Picture_5.jpeg)

![](_page_23_Picture_0.jpeg)

#### References

[1] J.J Witt, 'Modelling, Estimation, and Compensation of Imbalances in Quadrature Transceivers', Stellenbosch University, 2011.

[2] L. Hars, "Frequency Response Compensation with DSP", in Streamlining Digital Signal Processing 2nd Edition. Hoboken, NJ: Wiley, 2011, ch. 39.

[3] V. Valimaki and T. I. Laakso, "Principles of fractional delay filters," 2000 IEEE International Conference on Acoustics, Speech, and Signal Processing. Proceedings (Cat. No.00CH37100), Istanbul, Turkey, 2000, pp. 3870-3873 vol.6.

[4] 'IQ Correction', MEP Newsletter 3, 2011. [Online]

Available:http://www.delmarnorth.com/microwave/requirements/IQGainPhaseCo rrection.pdf, [Accessed: Aug. 6, 2018]

![](_page_23_Picture_7.jpeg)